
Digital Signal Processing (DSP)

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DIGITAL SIGNAL PROCESSING (DSP)

Lecture 3

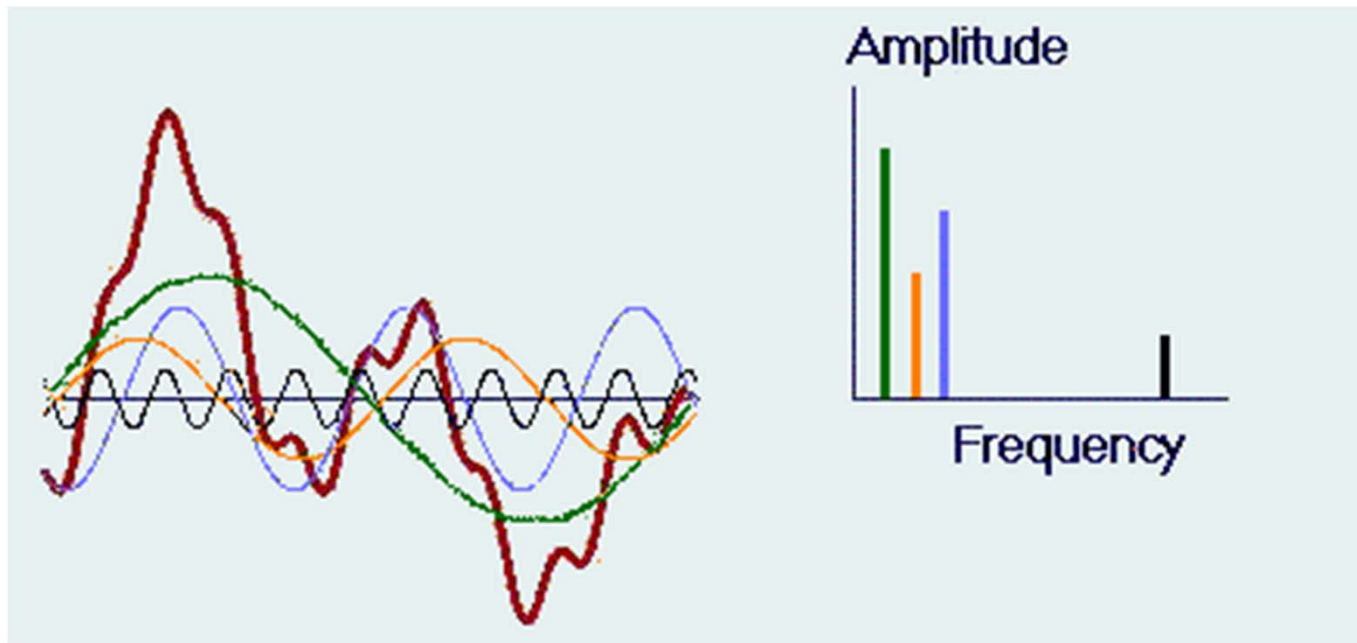
Discrete-time Fourier Transform (DTFT)

DTFT

- Frequency analysis vs. time analysis
 - An alternative tool
 - Equal information content
 - Fourier series, CT Fourier transform
 - Joseph Fourier: a French teacher
 - Expansion using the $\sin(\cdot)$ & $\cos(\cdot)$
 - Single tone: frequency content

DTFT

- Example: sum of sinusoidal signals



<http://www.qsl.net/on7yd/136narro.htm>

DTFT

- Interpretation of the sum: difficult in time domain
 - An easy task in frequency domain
 - Delta functions corresponding to each $\sin(\cdot)$
 - Determine frequency of $\sin(\cdot)$
 - Determine amplitude of $\sin(\cdot)$
 - Bolding some features!!
- A useful tool for LTI system analysis

DTFT

- Definition

- A transform that maps the DT signal $x[n]$ into the following continuous function (spectrum):

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}$$

- We use the notation $X(e^{j\omega})$ to emphasize on periodicity w.r.t. ω
- Generally complex-valued $X(e^{j\omega})$

DTFT

- Note

$$X(e^{j\omega}) = X_R(e^{j\omega}) + jX_I(e^{j\omega})$$

– Or

$$X(e^{j\omega}) = |X(e^{j\omega})|e^{j\angle X(e^{j\omega})}$$

– (X), dB: $20\log(X)$

– and

$$X(e^{j\omega+2\pi}) = X(e^{j\omega})$$

DTFT

- Inverse DTFT

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega,$$

- Observe the periodicity
- The frequency content interpretation
- DTFT pair: analysis & synthesis

$$\{x[n]\} \leftrightarrow X(e^{j\omega})$$

DTFT

- Example: $\{x[n]\} = \{a^n u[n]\}$
 - by definition,

$$\begin{aligned} X(e^{j\omega}) &= \sum_{n=0}^{\infty} e^{-j\omega n} \\ &= \sum_{n=0}^{\infty} (ae^{-j\omega})^n \\ &= \frac{1}{1 - ae^{-j\omega}} \end{aligned}$$

– exists when $|ae^{-j\omega n}| < 1$ i.e. $|a| < 1$

DTFT

- Convergence: an infinite sum!!
- Define the partial sum

$$X_M(e^{j\omega}) = \sum_{n=-M}^M x[n]e^{-j\omega n}$$

- Case 1(absolutely summable signals) the partial sum converges to the spectrum
- Case 2(energy signals): DTFT exists (but with a weaker statement)

DTFT

- Example (case 1): absolutely summability of the previous example

$$\sum_{n=-\infty}^{\infty} |x[n]| = \sum_{n=0}^{\infty} |a|^n = \frac{1}{1-|a|} < +\infty$$

- if $|a| < 1$

- Example (case 2-energy signals): Gibbs phenomenon

$$h_p[n] = \frac{\sin \omega_c n}{\pi n}$$

DTFT

- The partial sum convergence

$$H_M(e^{j\omega}) = \sum_{n=-M}^M \frac{\sin \omega_c n}{\pi n} e^{-j\omega n}$$

- The limit of the partial sum

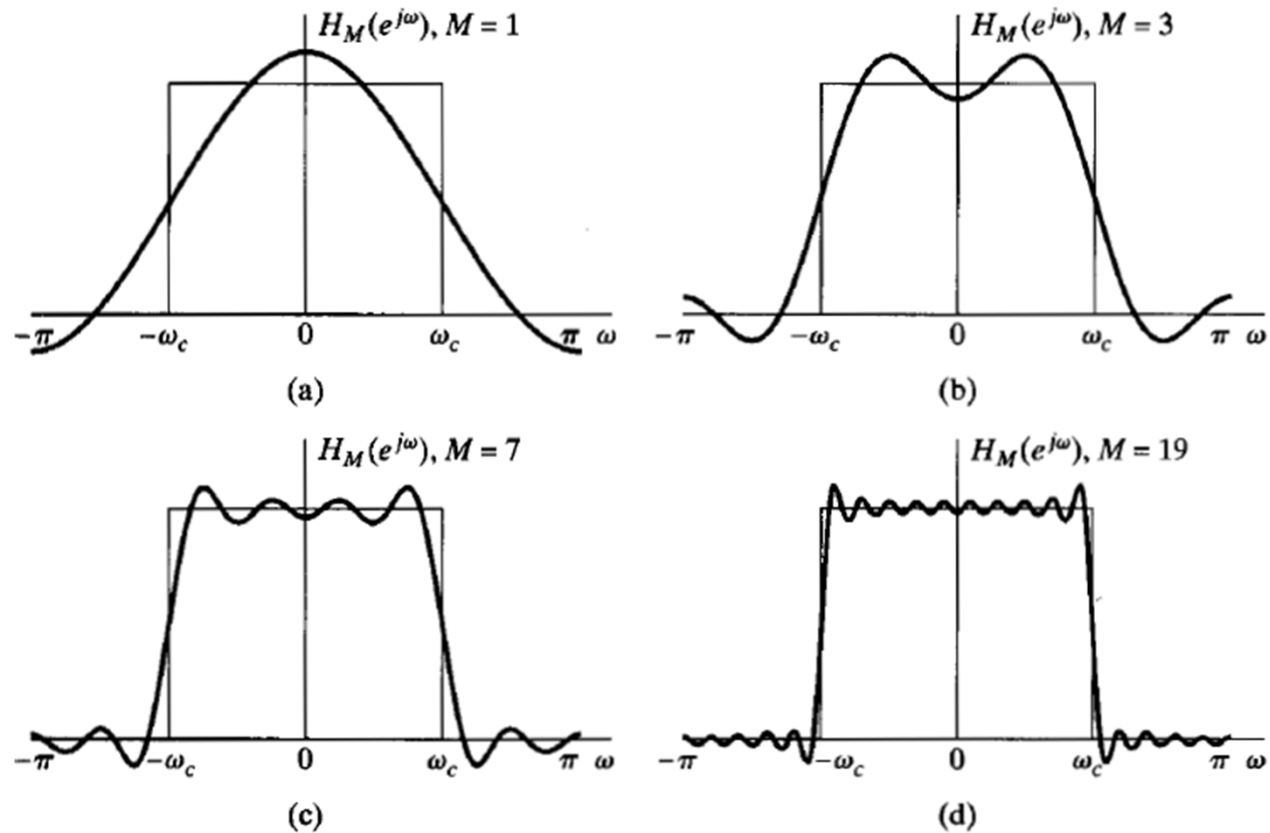
$$H_{lp}(e^{j\omega}) = \begin{cases} 1, & |\omega| < \omega_c, \\ 0, & \omega_c < |\omega| \leq \pi \end{cases}$$

- Indeed,

$$\lim_{M \rightarrow \infty} \int_{-\pi}^{\pi} |H_{lp}(e^{j\omega}) - H_M(e^{j\omega})|^2 d\omega = 0.$$

DTFT

– Illustration: Gibbs



DTFT

– Case 3(power signals): may converge

- The results contain delta functions

$$(a) \int_{-\infty}^{\infty} \delta(\omega) d\omega = 1$$

$$(b) \int_{-\infty}^{\infty} X(e^{j\omega}) \delta(\omega - \omega_0) d\omega = X(e^{j\omega_0}) \text{ if } X(e^{j\omega}) \text{ is continuous at } \omega = \omega_0;$$

$$(c) X(e^{j\omega}) \delta(\omega) = X(e^{j0}) \delta(\omega) \text{ if } X(e^{j\omega}) \text{ is continuous at } \omega = 0$$

- Note also that $\delta(\omega) = 0$ for $\omega \neq 0$

DTFT

- Example (DTFT of a constant): $x[n] = 1$ for all n
 - Claim

$$X(e^{j\omega}) = \sum_{k=-\infty}^{\infty} 2\pi\delta(\omega + 2\pi k)$$

- Proof via the synthesis equation

$$\begin{aligned} x[n] &= \int_{-\pi}^{\pi} \sum_{h=-\infty}^{\infty} \delta(\omega + 2\pi h) e^{j\omega n} d\omega \\ &= \int_{-\pi}^{\pi} \delta(\omega) e^{j\omega n} d\omega = 1 \end{aligned}$$

DTFT

- Example (DTFT of CES): $x[n] = e^{j\omega_0 n}$

– Claim

$$X(e^{j\omega}) = \sum_{k=-\infty}^{\infty} 2\pi \delta(\omega - \omega_0 + 2\pi k), \quad -\pi < \omega_0 \leq \pi$$

– Similar proof

$$\begin{aligned} x[n] &= \frac{1}{2\pi} \int_{-\pi}^{\pi} \sum_{h=-\infty}^{\infty} 2\pi \delta(\omega - \omega_0 + 2\pi h) e^{j\omega n} d\omega \\ &= \int_{-\pi}^{\pi} \delta(\omega - \omega_0) e^{j\omega n} d\omega \end{aligned}$$

DTFT

- Properties of DTFT

- Periodicity

As noted earlier that the DTFT $X(e^{j\omega})$ is a periodic function of ω with period 2π . This property is different from the continuous time Fourier transform of a signal.

- Linearity

If $\{x[n]\} \leftrightarrow X(e^{j\omega})$
and $\{y[n]\} \leftrightarrow Y(e^{j\omega})$
then $a\{x[n]\} + b\{y[n]\} \leftrightarrow aX(e^{j\omega}) + bY(e^{j\omega})$

DTFT

– Conjugation

$$\{x^*[n]\} \leftrightarrow X^*(e^{-j\omega})$$

- A simple proof

$$\begin{aligned} \sum_{n=-\infty}^{\infty} x^*[n]e^{-j\omega n} &= \sum_{n=-\infty}^{\infty} [x[n]e^{j\omega n}]^* \\ &= \left[\sum_{n=-\infty}^{\infty} x[n]e^{-j(-\omega)n} \right]^* \\ &= X^*(e^{-j\omega}) \end{aligned}$$

– Time reversal

$$\{x[-n]\} \leftrightarrow X(e^{-j\omega})$$

DTFT

- Time shifting

$$\{x[n - n_0]\} \leftrightarrow e^{-j\omega n_0} X(e^{j\omega})$$

- Modulation

$$\{e^{j\omega_0 n} x[n]\} \leftrightarrow X(e^{j(\omega - \omega_0)})$$

DTFT

- Symmetry property (Hermitian symmetry)
 - Let $x[n]$ is real, then

$$x[n] = x^*[n]$$

- which leads to the following symmetry

$$X(e^{j\omega}) = X^*(e^{-j\omega})$$

- Proof : straightforward
- Useful/important results